





**DNA 5442T** 

# SUITABILITY OF ARES FOR SIMULATING TACTICAL BURST EMP ENVIRONMENTS

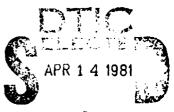
D. W. Holst
Mission Research Corporation
P.O. Drawer 719
Santa Barbara, California 93101

1 April 1980

Topical Report for Period 15 January 1980-1 April 1980

CONTRACT No. DNA 001-80-C-0076

APPROVED FOR PUBLIC RELEASE; DISTRIBUTION UNLIMITED.



A

THIS WORK SPONSORED BY THE DEFENSE NUCLEAR AGENCY UNDER RDT&E RMSS CODE B323080464 X99QAXVC30102 H2590D.

Prepared for

Director

**DEFENSE NUCLEAR AGENCY** 

Washington, D. C. 20305

81 4 14 V<sub>0</sub>8

Destroy this report when it is no longer needed. Do not return to sender.

PLEASE NOTIFY THE DEFENSE NUCLEAR AGENCY, ATTN: STTI, WASHINGTON, D.C. 20305, IF YOUR ADDRESS IS INCORRECT, IF YOU WISH TO BE DELETED FROM THE DISTRIBUTION LIST, OR IF THE ADDRESSEE IS NO LONGER EMPLOYED BY YOUR ORGANIZATION.

SECURITY CLASSIFICATION OF THIS PAGE (When Date Entered)

REPORT DOCUMENTATION		READ INSTRUCTIONS BEFORE COMPLETING FORM
DNA 5442T	AD-AC47 70	3. RECIPIENT'S CATALOG NUMBER
SUITABILITY OF ARES FOR SIMULATING BURST EMP ENVIRONMENTS.	\ \	Topical Report Feriod Covered  Topical Report for Period  15 Jan 1 Apr 80 1
7 AUTHORIA	14	MRC-R-565
D. W. Holst		DNA 001-80-C-0076
9 PERFORMING ORGANIZATION NAME AND ADDRESS Mission Research Corporation  P.O. Drawer 719 Santa Barbara, California 93102		O PROGRAM ELEMENT PROJECT TASK AREA WORK UNIT NUMBERS SUDEASK X990AXVC301-02
11 CONTROLLING OFFICE NAME AND ADDRESS Director Defense Nuclear Agency Washington, D.C. 20305 14 MONITORING AGENCY NAME & ADDRESS(II different	t from Controlling Office)	12 REPORT DATE  1 Apr 180  13 NUMBER OF PAGES  72  15 SECURITY CLASS (of this report)
		UNCLASSIFIED  15a DECLASSIFICATION DOWNGRADING SCHEDULE
17 DISTRIBUTION STATEMENT (of the abstract entered )	in Block 20, if different from	n Report)
18 SUPPLEMENTARY NOTES		
This work sponsored by the Defens B323080464 X99QAXVC30102 H2590D.	e Nuclear Agency	under RDT&E RMSS Code
19 KEY WORDS (Continue on reverse side if no ressary and	d identify by block number)	<del></del>
ARES Source Region Coupling	Simulation Tactical System EMP	
The possibility of using the ARES tactical systems is discussed. A environment at the inner edge of burst at a range of 1.2 km. The ARES pulse are compared with those of the ARES and the source region small loop are compared. These of	to evaluate sound tactical sour this source region time and amplitue from the selectields into a s	ce region is defined and the on is determined for a 50 kt de characteristics of the ted environment. Coupling hort monopole antenna and

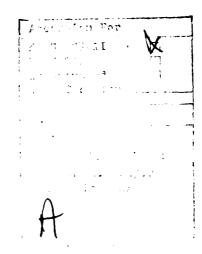
DD 1 JAN 73 1473 EDITION OF 1 NOV 65 IS OBSOLETE

UNCLASSIFIED

SECURITY CLASSIFICATION OF THIS PAGE (When Data Entered)
20. ABSTRACT (Continued)
in tactical systems. Based upon these results suggestions are made for augmenting and supplementing testing in the ARES to obtain responses representative of the source region.

#### PREFACE

The author wishes to express thanks to Dr. William Crevier for providing much of the source region data and for many helpful discussions throughout the course of this work. Mr. Robert Hamilton provided the Fourier transforms and time derivatives for the source region and ARES data.



### TABLE UF CONTENTS

SECTION		PAGE
	PREFACE	1
	LIST OF ILLUSTRATIONS	3
1	INTRODUCTION	5
2	THE SOURCE REGION ENVIRONMENT	10
	2.1 TYPICAL ENVIRONMENT	11
	2.2 ARES EMP ENVIRONMENT	21
3	SOURCE REGION COUPLING	34
	3.1 COUPLING MODIFICATION IN THE SOURCE REGION	l
	ENVIRONMENT	30
	3.2 COUPLING TO CANONICAL OBJECTS 3.2.1 Short Monopole Antenna 3.2.2 Small Loop Antenna	39 39 45
4	ARES MODIFICATIONS FOR SOURCE REGION TESTS	52
5	SUDMARY AND CONCLUSIONS	5/
	REFERENCES	62

## LIST OF ILLUSTRATIONS

FIGURE		PAGE
1-1	ARES facility description.	7
1-2	Coordinate system for ARES.	8
2-1	Variation of the peak Compton current $\mathbf{J}_{\mathbf{r}}$ with range for a 50 kt surface burst.	12
2-2	Variation of the peak electric and magnetic fields with range for a 50 kt surface burst.	13
2-3	Variation of the peak air conductivity $\boldsymbol{\sigma}$ with range for a 50 kt surface burst.	14
2-4	Radial Compton current vs time at 1.2 km from a 50 kt surface burst.	16
2-5	Electric and magnetic fields vs time at 1.2 km from a 50 kt surface burst.	18
2-6	Air conductivity vs time at 1.2 km from a 50 kt surface burst.	19
2-7	Source region displacement current and conduction current densities at 1.2 km from a 50 kt surface burst.	20
2-8	Comparison of horizontal magnetic fields for the ARES and source region.	23
2-9	Comparison of vertical electric fields for the ARES and source region.	24
2-10	Comparison of longitudinal electric fields for the ARES and source region.	25
2-11	Time derivative of the vertical electric field for the ARES and source region showing relative magnitudes.	27
2-12	Time derivative of the horizontal magnetic field for the ARES and source region showing relative magnitudes.	28

## LIST OF ILLUSTRATIONS (continued)

FIGUE	RE.	PAGE
2-13	Fourier transform of magnetic field for ARES and for source region.	29
2-14	Fourier transform of vertical electric field for ARES and for source region.	30
2-15	Fourier transform of longitudinal electrtic field for ARES and for source region.	31
2-16	Conduction current and displacement current densities showing relative magnitudes.	33
3-1	Typical coupling elements for deployed tactical systems.	35
3-2	Equivalent circuit for short monopole in the source region.	41
3-3	Response of a short antenna 1.2 km from a 50 kt surface burst.	44
3-4	Response of a short antenna 2.0 km from a 50 kt surface burst.	46
3-5	Comparison of short antenna response in ARES and source region environments.	47
3-6	Equivalent circuit for small loop in the source region.	49
3-7	Comparison of small loop response in ARES and source region environments.	50

# SECTION 1 INTRODUCTION

The ARES (Advanced Research Electromagnetic Simulator) is a parallel-plate transmission line simulator which provides a working volume 33 meters long by 40 meters wide by 40 meters high. It was originally designed and constructed to provide a free space (plane wave) EMP environment for the testing of full scale missiles in simulated flight.

This report considers application of the ARES to source region testing near the air-ground interface. Since the ARES was designed with a very different problem in mind, it is reasonable to anticipate that this new application will require some modification of the facility. Before the nature of these changes, if any, can be defined we must first determine to what extent the present ARES configuration compares with a typical source region environment. Some elements of the source region environment, such as air conductivity, will be missing entirely. Other parameters may be present but of incorrect sense or amplitude. The impact of the missing or modified parameters must be assessed. For the purposes of this report it was deemed best to choose a "typical" source region environment. This approach allows specific comparisons to be made, and can result in specific recommendations for modifications.

A major attribute of the ARES is its large physical size, allowing for the possibility of testing large tactical systems or components - vans, vehicles, towers, with their numerous cable interconnections.

Figure 1-1 provides a schematic representation of the ARES facility. All dimensions are in meters. The working volume is accessable via a surface road so that large vehicles and system components have direct access to the working floor space. Figure 1-2 shows the coordinate system employed for past field maps of the ARES working volume. The same coordinate system will be employed in this report to facilitate comparisons with previous work.

While a fully deployed tactical system covers considerably more territory than available in the ARES, major elements of a system could conceivably be tested under deployed conditions. For example, a communications and control van, a missile launcher, and their associated cabling and/or antennas might be arranged in the available ARES floor space to simulate a realistic deployment. Due to this unique capability the balance of this report will place emphasis on large tactical systems in the source region environment. Smaller systems or components can be tested in other simulators, such as AURORA, which are already operated as source region simulators but lack the large volume capability.

A previous study of the ARES as a source region simulator has been conducted by HDL $^1$ . In that study, the ARES is accepted for what it is, a plane wave simulator, and the comparison of the ARES and the source region environments is conducted at a range sufficiently far from the burst point that  $E_\theta$  and  $B_\varphi$  have their free field ratio of the speed of light, or equivalently, that the air conductivity is small ( $\sigma < 10^{-6} \, \mathrm{mho/m}$ ).

It is felt that the above aproach may be overly-restrictive and in this report we adopt a different viewpoint. Rather than concentrating on the outer edge of the source region, we will concentrate on the inner

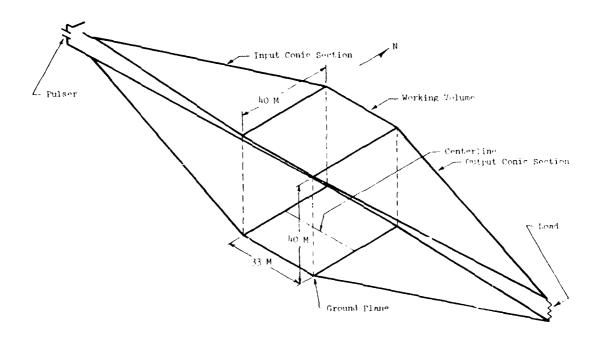


Figure 1-1. ARES facility description.

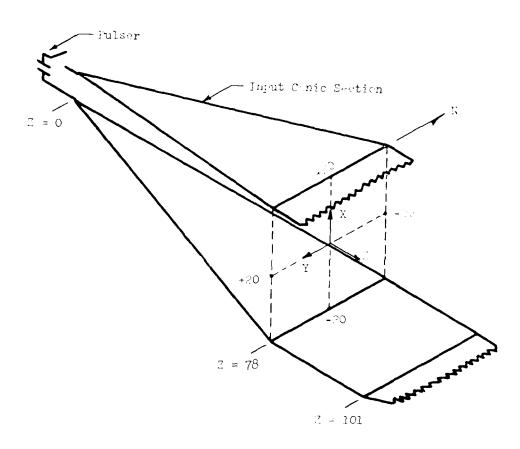


Figure 1-2. Coordinate system for ARES.

edge of the tactical source region - the minimum range at which crew survival is insured. Then equipment survival is also necessary for mission completion. This is the range for which our environment comparisons and coupling calculations were conducted. The EMP fields at this range are not plane wave, nor are the air conductivity and Compton currents insignificant. However, a plane wave simulator may still provide useful test results and this possibility should be examined in greater detail.

The comparisons described in later sections were conducted for a device with a  $50~\rm kT$  yield, with 3 percent of the yield in immediate neutron and gamma radiation. This is felt to be a reasonable upper limit for tactical nuclear weapons. The immediate dose from this device determines the crew survival range.

# SECTION 2 THE SOURCE REGION ENVIRONMENT

Before proceeding with the technical discussion, we need to define what constitutes the "source region of interest". This is partly a philosophical and partly a physical definition since it must be guided by the prevailing battlefield philosophy as well as by the laws of physics.

The basic Army philosophy regarding nuclear survivability is that the particular item of equipment or system should survive if a sufficient percentage of its crew survives long enough to complete the mission<sup>2</sup>. Equipment failures can be generated by several nuclear weapon effects such as air blasts, thermal radiation, and EmP. In this report we will be concerned with how one verifies experimentally that an army tactical system is hard to source region EMP. The air blast and thermal radiation, as well as total radiation dose, serve to define an inner radius to the source region since they would disable any crew located nearer the burst point.

The definition of the inner edge of the source region is somewhat arbitrary. It depends upon the type of protection a system provides the crew from nuclear weapon effects. The Army specifically considers: exposed personnel, personnel in foxholes, personnel in armored vehicles, and personnel in wheeled vehicles. It also depends upon the weapon output to which one expects to be exposed. The relative magnitudes of the various disabling mechanisms can vary greatly with weapon design.

In addition the EMP is sensitive to such parameters as the water content of the air, soil conductivity, and terrain features such as hills and valleys. It can also be modified by previous nuclear explosions.

#### 2.1 TYPICAL ENVIRONMENT

In spite of the variability in the source region EMP 't is useful to define a specific environment to which one may refer. Therefore, let us consider 50 kT yield as an upper limit for weapons of tactical interest. Assume that 3 percent of the yield is in initial neutron and  $\gamma$ -radiation. A typical mean free path ( $\lambda$ ) at sea level for such radiation is  $3\times10^4$  cm. Using the fact that 1 kT equals  $4.18\times10^{19}$  ergs, 1 rad = 100 ergs/gm, and the density of air is  $\rho$  =  $1.23\times10^{-3}$  gm/cm<sup>3</sup>, we find that the immediate dose at a radius r is

$$D = \frac{Y \times 0.03 \times 4.18 \times 10^{19} e^{-r/\lambda}}{100 \times 4\pi r^2 \times \rho \lambda}$$

$$D = 1.35 \times 10^{15} e^{-r/\lambda}/r^2$$
 rads

At  $r = 10^5$  cm, D = 4800 rads, while at  $r = 1.5 \times 10^5$ , D = 405 rads. Doses in the 1000 to 5000 rad range produce almost immediate incapacitation<sup>3</sup> so we may define the inner radius of the source region to be somewhere in the 1 to 1.5 km range.

Figures 2-1 through 2-3 display the variation in the peak values of the radial Compton current,  $J_r$ , the radial electric field,  $E_r$ , the azimuthal magnetic field,  $B_{\phi}$ , the vertical electric field,  $E_{\theta}$ , and the air conductivity,  $\sigma$ , with range for a 50 kT surface burst. All values are

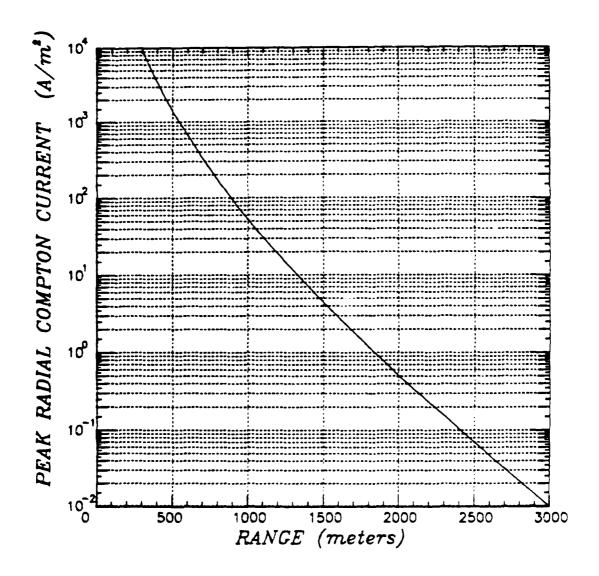


Figure 2-1. Variation of the peak Compton current  $\, J_r \,$  with range for a 50 kt surface burst. The Compton current is in the negative r-direction.

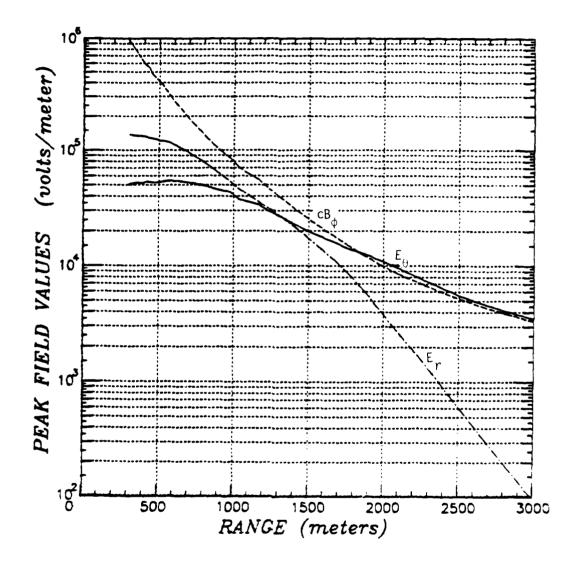


Figure 2-2. Variation of the peak electric ( $E_r$ ,- $E_\theta$ ) and magnetic (- $cB_\phi$ ) fields with range for a 50 kt surface burst.

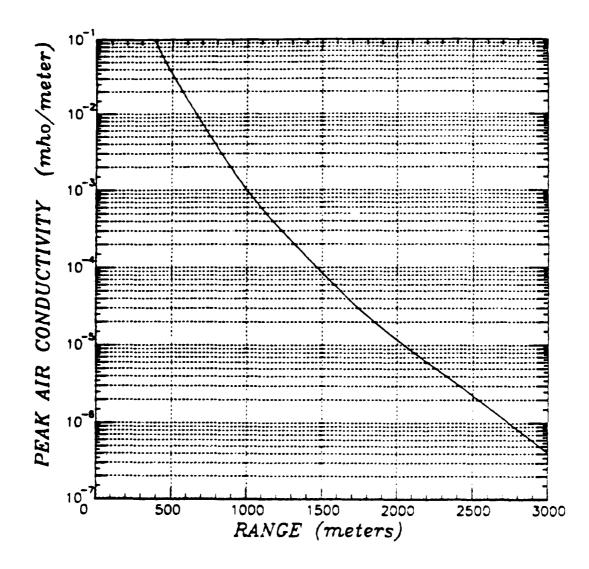


Figure 2-3. Variation of the peak air conductivity with range for a 50 kt surface burst.

near the ground-air interface. Results are from the new MRC/HDL code MODELC which assumes a perfectly conducting ground. A one percent air moisture content was assumed.

Figure 2-1 shows that in the 1 to 1.5 km range of interest the radial current decreases from 55 to 4.5 amps/m². The resulting radial electric field  $E_{\Gamma}$  varies from 50 to 18 kV/meter. The vertical gradient in the radial electric field at the ground results in the generation of  $B_{\varphi}$  and  $E_{\theta}$  field components. It is seen that  $B_{\varphi}$  varies from about  $2.8\times 10^{-4}$  to about  $10^{-4}$  Tesla over the ranges of interest. The magnetic field is plotted as  $cB_{\varphi}$  (in kV/m) in Figure 2-2 to emphasize the distinction between the source region and plane wave propagation. The fields become essentially plane wave when  $E_{\theta}=cB_{\varphi}$  at about 1.8 km. The vertical electric field shows less variation ranging from about 40 to 20 kV/meter. The air conductivity changes from  $10^{-3}$  to  $10^{-4}$  mho/meter over the 1.0 to 1.5 km range.

For the following discussion we select a range of  $1.2~\rm km$  from the burst point. The immediate dose at this range is estimated to be  $1720~\rm km$  rads. We therefore proceed to define a specific environment under the following assumptions:

- (a) Surface burst
- (b) 50 kT yield
- (c) Range from burst = 1.2 km
- (d) Perfectly conducting ground
- (e) Air moisture content = 1 percent
- (f) Air density =  $1.23 \times 10^{-3}$  gm/cm<sup>3</sup>

The EMP environment parameters at our designated range are defined by Figures 2-4 through 2-6 which present computed time histories. The Compton current, shown in Figure 2-4, has an initial e-folding time of  $10^{-8}$  sec.

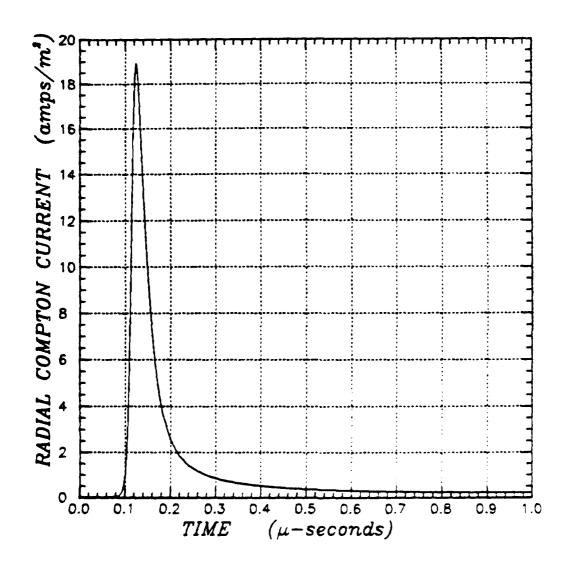


Figure 2-4. Radial Compton current versus time at 1.2 km from a 50 kt surface burst. The radial component of the Compton current is negative.

At  $4\times10^{-8}$  sec it decreases to  $5\times10^{-9}$  seconds. The peak current is about 19 A/m which occurs at  $1.2\times10^{-7}$  seconds.

The assumption of a perfectly conducting ground causes the radial electric field to be shorted out so that  $E_r=0$  there. However, at relatively short distances above the ground ( $\simeq$  3 meters)  $E_r$  is not zero but is given in Figure 2-5. The peak value is about 34 kV/meter, dropping to a relatively constant value of 15 kV/meter for later times.

Figure 2-5 shows that the surface magnetic field continues to rise beyond the peak in the Compton current. It eventually reaches a maximum value of about  $1.8 \times 10^{-4}$  Tesla which corresponds to an H of 143 A/m. It should be noted that  $B_{\varphi}$  is negative. This corresponds to a positive surface current that is flowing radially outward or an electron flow back towards the burst point.

The vertical electric field in Figure 2-5 points upward at the surface which corresponds to a negative  $E_{\theta}$ . At the peak of the Compton current  $E_{\theta}$  also peaks. It continues decreasing until  $4\times10^{-7}$  seconds when it begins increasing again. By 10  $_{\mu}$ seconds it has increased to nearly the value it had at the peak of the prompt  $_{\gamma}$ -pulse.

Air conductivity versus time is shown in Figure 2-6. The air conductivity peaks at about  $3.5\times10^{-4}$  mho/m, just after the peak in the Compton current. By 10 µseconds it has fallen to about  $10^{-5}$  mho/m. Because of the air conductivity both conduction ( $\sigma E$ ) and displacement ( $\epsilon_0\dot{E}$ ) current densities are present in the source region. These two currents are presented in Figure 2-7 with the displacement current labeled D (solid curve) and the conduction current labled C (dashed curve). They are seen to dominate at different times, reaching comparable peak magnitudes.

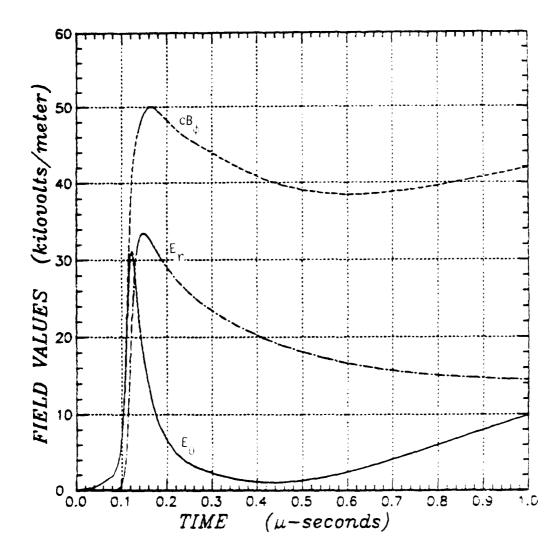


Figure 2-5. Electric ( $E_r$ ,- $E_r$ ) and magnetic (-cB<sub>r</sub>) fields versus time at 1.2 km from a 50 kt surface burst. Note that the scale is in kilovolts/meter.

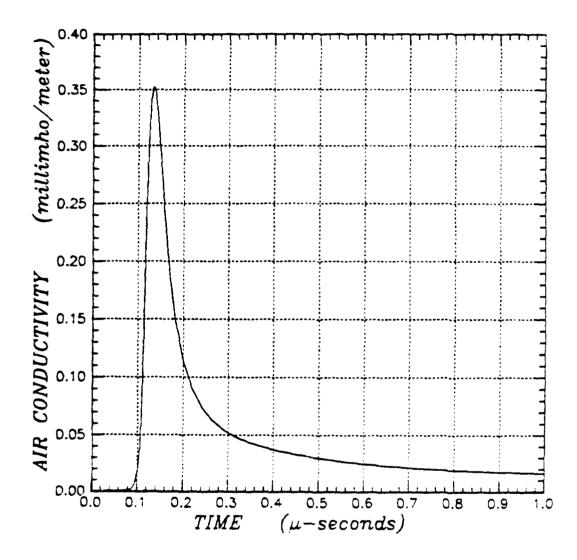


Figure 2-6. Air conductivity versus time at 1.2 km from a 50 kt syrface burst. Note that the scale is in millimho/meter.

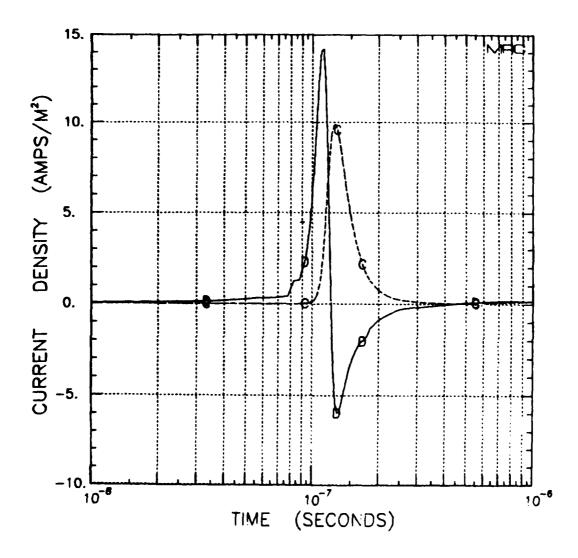


Figure 2-7. Source region displacement current ( $-\epsilon_0\dot{E}_\theta$ ) and conduction current ( $-\sigma E_\theta$ ) densities at 1.2 km from a 50 kt surface burst.

It should be noted that the ARES provides the correct fields in terms of components present and sense, though amplitudes differ. For example, the ARES  $E_X$  corresponds to the source region  $-E_\theta$  since the source region vertical electric field is negative in the spherical coordinate system (points upward). Similarly, we have  $E_Z \approx -E_R$  and  $B_y = -B_{\varphi}$ . Because of this one-for-one correspondence, the ARES field designations  $E_X$ ,  $B_Y$ ,  $E_Z$  (see Figure 1-2) will be adopted throughout the remainder of this report.

Adaptation of the ARES as a source region simulator must be examined in terms of the source region environment generated by a nuclear surface burst. Physical descriptions of the source region, which elaborate upon the characteristics described above, may be found in References 4 and 5.

#### 2.2 ARES EMP ENVIRONMENT

In the previous section the source region environment 1.2 km from a 50 kt burst has been defined in terms of the time histories of several source and field parameters. Comparisons between the ARES and the source region defined above have been conducted for several environment parameters with the results summarized in Figures 2-8 to 2-16.

ARES data were obtained from References 6 through 9. Basically, the ARES is a parallel plate transmission line simulator with conical input and output sections. It operates in the TEM mode, with  $\rm E_{\rm X}$  and  $\rm B_{\rm Y}$  the principal electric and magnetic field components respectively. Due to spherical wave, diffraction and reflection effects a longitudinal (E<sub>Z</sub>) electric field component is present and the propagation velocity is about 8 percent below the velocity of light (the wave impedance is not quite plane wave). The ARES coordinate system is shown in Figure 1-2.

For the parameter comparisons presented below the time plot for the measured ARES magnetic field,  $B_y$ , was employed. The  $E_x$  and  $E_z$  electric field time variations were assumed to be the same with amplitude adjustments to fit measured peak values.

Comparisons of the pulse time histories for the ARES and the source region are presented in Figures 2-8 to 2-10. The ARES is shown as the solid line with a dashed line for the source region. The magnitude of  $E_{\rm X}$  in the ARES exceeds the source region amplitude by about 70 percent. The peak magnitude of  $E_{\rm Z}$  in the ARES is only about 30 percent of the source region value but the peak amplitudes of the  $E_{\rm Y}$ 's are nearly identical.

The ARES pulser is capable of providing a very fast rise time with typical values between 5 and 10 ns. The source region pulse shows an e-folding with  $\alpha\approx2\times10^8/\text{sec}$ . This parameter depends upon the device, burst altitude and ground conductivity but the value shown here is not untypical. With a 10-90 percent rise time of 10 ns the ARES provides a very good simulation of the leading edge of the source region pulse.

For late times the source region fields maintain high amplitudes, except for  $E_X$  which first shows a rapid drop after the peak before again climbing to its peak value. By  $10^{-3}$  seconds all field components have decreased to small values. The ARES fields drop rapidly after the peak, providing a reasonably good simulation of the early time vertical electric field but showing no correspondence to the field components for late times. It will be seen in the next section, however, that the coupling response is often more sensitive to E and B, the time derivatives of the electric and magnetic fields, than to the field components proper. We should therefore examine the time derivatives to evaluate their impact on the source region simulation.

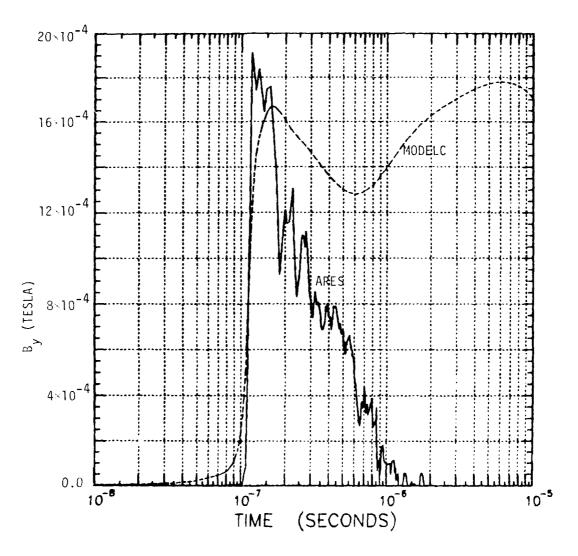


Figure 2-8. Comparison of horizontal magnetic fields for the ARES and source region.

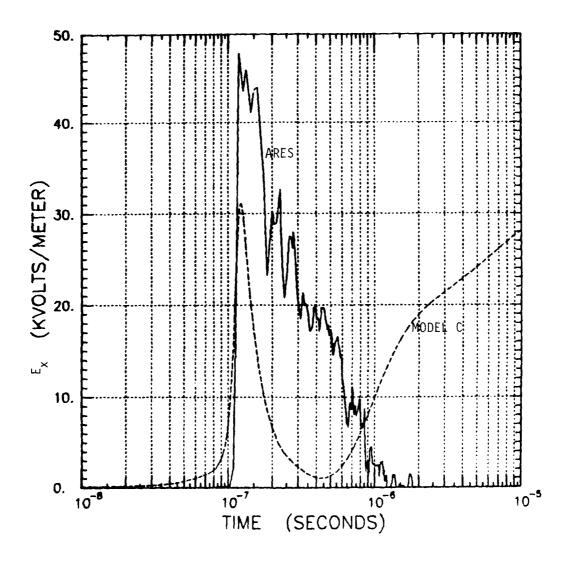


Figure 2-9. Comparison of vertical electric fields for the ARES and source region. Note that the scale is kilovolts/meter.

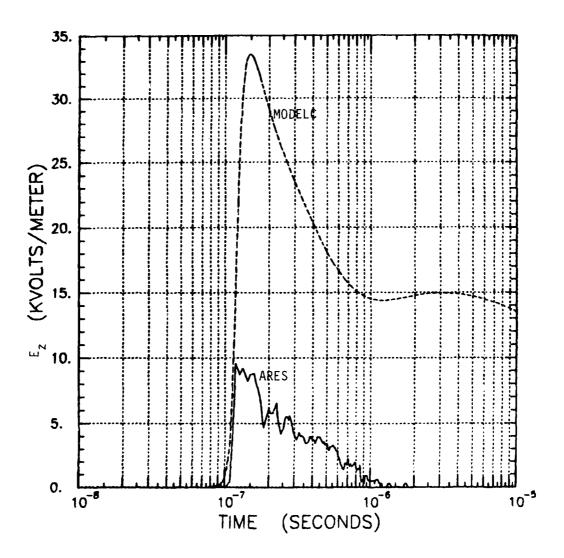


Figure 2-10. Comparison of longitudinal electric fields for the ARES and source region. Note that the scale is kilovolts/meter.

Time derivatives of  $E_X$  and  $B_y$  are presented in Figures 2-11 and 2-12, shown by a dashed line. They were computed using the source region results from the MODELC code. The ARES results are presented on each plot as a solid line. It is seen that the magnitudes for the ARES exceed the source region values by a factor of six for both  $\tilde{E}_X$  and  $\tilde{B}_y$  providing a significant margin of overtest capability. Since dipoles and loops respond primarily to  $\tilde{E}_X$  and  $\tilde{B}_y$  respectively, the figures show the relative amplitudes of drives for these elements in the ARES and source region. For the  $\tilde{E}_X$  negative overshoot, the ARES exceeds the MODELC value by nearly a factor of 2.

Fourier transforms of all field components are presented in Figures 2-13 to 2-15. In each figure the ARES is represented by a solid line and the source region by a dashed line. The ARES transforms were obtained from the time histories shown in Figures 2-8 to 2-10. They show the characteristic notches at 17 and 34 MHz and in general are the same as transforms shown in the references.  $^{7,9}$  For all field components the ARES shows higher amplitudes at high frequencies and lower amplitudes at low frequencies. It will therefore provide an overtest capability at high frequencies where many system components show the greatest coupling response.

Since the source region field amplitudes generally remain high to late times the transforms show higher amplitudes at low frequencies. This is particularly true for the radial electric field  $\mathbf{E}_Z$  which has a large amplitude in the source region (see Figure 2-10). The net result is that the ARES provides an overtest at high frequencies and an undertest at low frequencies. The crossover frequency between undertest and overtest is different for each field component, being 2 MHz for the horizontal magnetic field, 180 KHz for the vertical electric field, and so MHz for

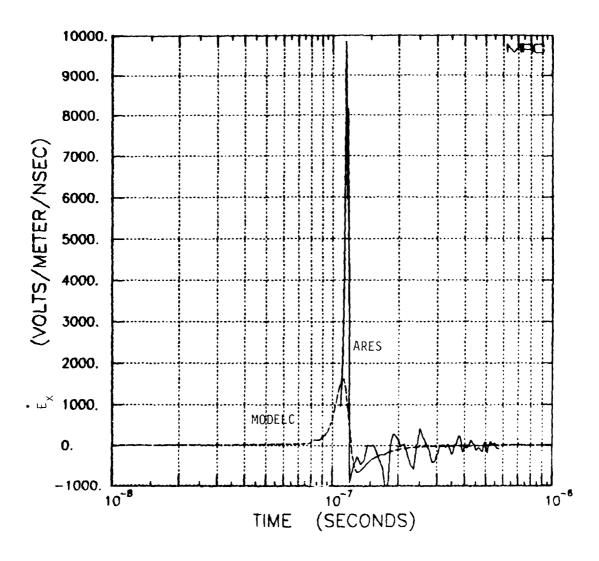


Figure 2-11. Time derivative of the vertical electric field  $(\dot{E}_\chi)$  for the ARES and source region showing relative magnitudes. Note that the vertical scale is volts/meter/nanosecond.

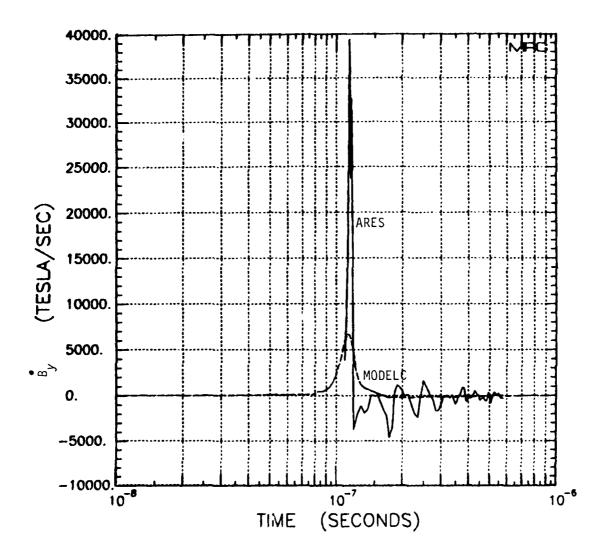


Figure 2-12. Time derivative of the horizontal magnetic field  $(\mathring{B}_y)$  for the ARES and source region showing relative magnitudes.

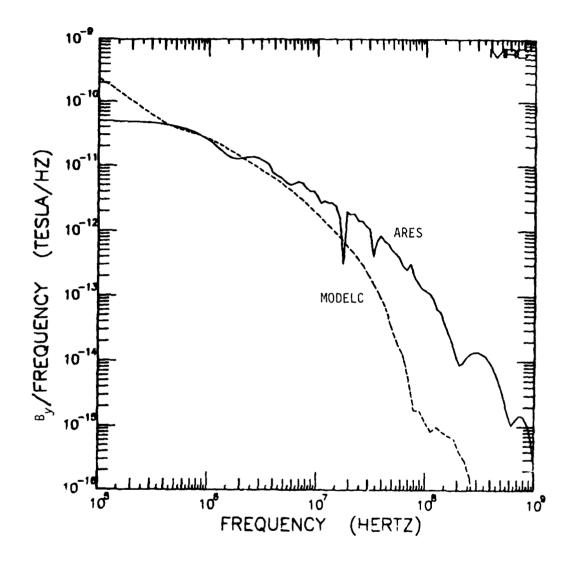


Figure 2-13. Fourier transform of magnetic field for ARES and for source region.

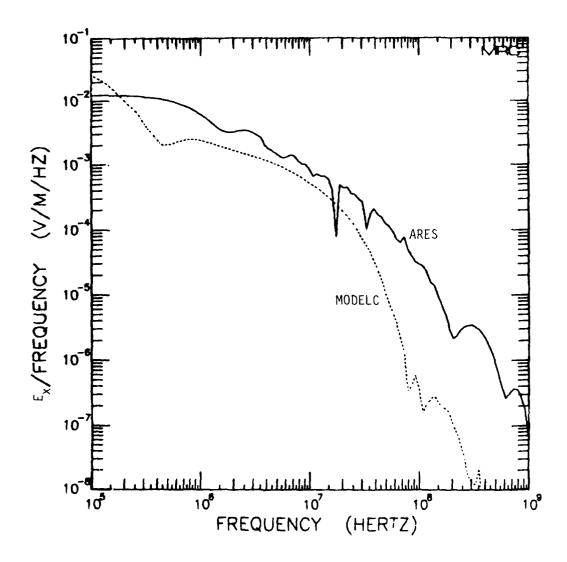


Figure 2-14. Fourier transform of vertical electric field for ARES and for source region.

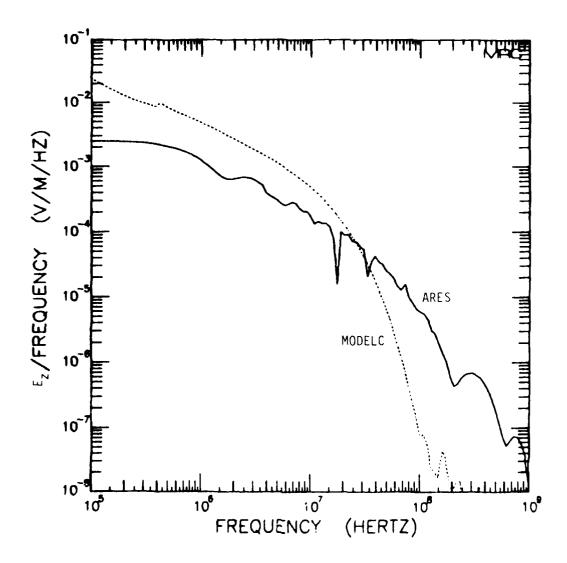


Figure 2-15. Fourier transform of longitudinal electric field for ARES and for source region.

the longitudinal electric field. It would be desirable to enhance the amplitude of the ARES  $\mathsf{E}_\mathsf{Z}$  component. A possible means for accomplishing this is described in Section 4.

The source region contains both conduction and displacement current densities which dominate at different times. Prior to the peak of the pulse the displacement current is dominant since  $\dot{E}_{\chi}$  is large and the increasing  $\sigma$  trails  $E_{\chi}$  in time. This gives

$$\epsilon_0 \dot{\epsilon}_x >> \sigma \epsilon_x$$
 (before peak)

At the peak  $E_x$  = 0 so the conduction current is dominant giving

$$\sigma E_{x} >> \varepsilon_{0}^{\epsilon} E_{x}$$
 (at peak).

For late times both  $\sigma$  and  $\tilde{E}_{\chi}$  are much smaller so that both the conduction current ( $\sigma E_{\chi}$ ) and the displacement current ( $\epsilon_0 \tilde{E}_{\chi}$ ) are small. They have comparable amplitudes at late times for the tactical environment selected, so

$$\sigma E_{x} \approx \varepsilon_{0} \dot{E}_{x}$$
 (late times)

For the ARES the conduction current is zero since the air conductivity  $\sigma = 0$ . However, the ARES provides a significant displacement current density (amps/m²) as shown in Figure 2-16. For comparison the source region conduction current (labeled C) and displacement current (labeled  $\sigma_f$  are shown by dashed lines. It is seen that the ARES provides an overtest by a factor of 6 for the displacement current density.

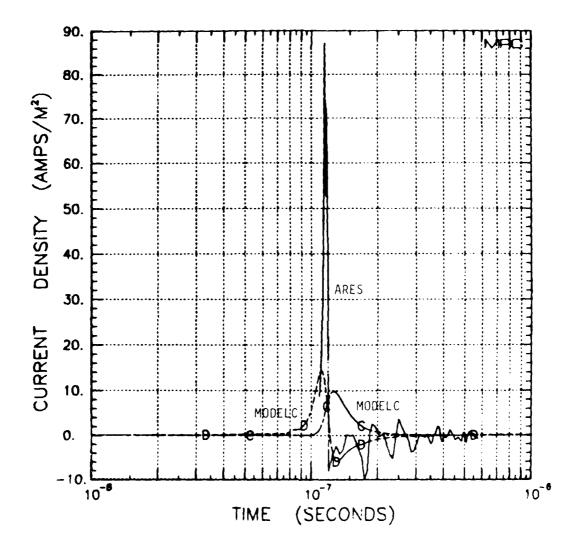


Figure 2-16. Conduction current ( $\cdot E_{\chi}$ ) and displacement current ( $\cdot e_{\chi}^{\dagger}$ ) densities showing relative magnitudes. Conduction current is zero in the ARES since  $\cdot = 0$ .

# SECTION 3 SOURCE REGION COUPLING

EMP coupling in the source region is characterized by a number of phenomena which are absent for plane wave coupling. These phenomena are described briefly below and their relative importance for the tactical source region (as defined in this report) is discussed.

The quantity and variety of coupling sources for a deployed tactical system is staggering; it is not possible to consider them all in this report. Broadly, however, many of the coupling mechanisms encountered may be separated into three general catagories which respond to different components of the source region environment. For the tactical system these are dipoles, loops and transmission lines. While these threecutagories do not exhaust the possibilities for coupling into a large system, they do represent three very important and common coupling mechanishs. Some typical examples are shown in Figure 3-1. Examples of dipoles (or monopoles) are whip antennas on vans, trucks, etc., towers supporting lines or other antennas, or virtually any vertical conductor. For loops we find actual loss antennas; small loops formed with ground cable, running between vans; and larger loops formed by two or more vehicles, the iround, and caldes strand above ground between the vehicles. For transmission line, we find the numerous power lables, phone Times, Reliables, etc., strung between all elements of the deployed rister on or above the ground.

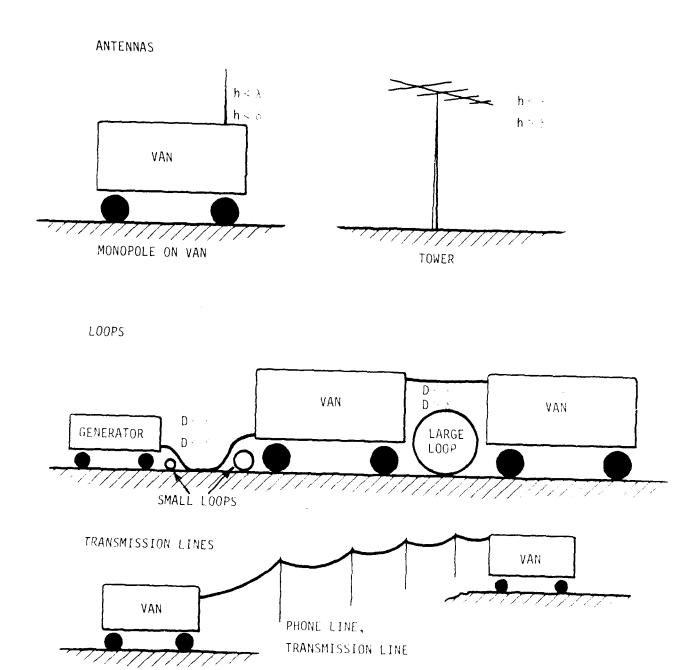


Figure 3-1. Typical coupling elements for deployed tactical systems.

We have elected to compute the response for typical examples of two of these components. These "canonical" objects are a short monopole over a ground plane and a small loop. These objects will display representative responses to the environment defined previously. The relative importance of some parameters, such as air conductivity, in determining the overall object response shall also be examined briefly.

### 3.1 COUPLING MODIFICATION IN THE SOURCE REGION ENVIRONMENT

Distinct from plane wave coupling, the source region introduces several new phenomena into the coupling picture. In this section we shall simply list the important parameters with a brief description of their impact on the general coupling environment. More detailed descriptions of source region coupling for specific objects are presented in the next sections. It is important to note that it is not necessary to accurately duplicate all aspects of the tactical environment. Rather, it is only necessary to reproduce those aspects of the environment which are important in determining the coupling response of the system. The importance of the various new parameters will be discussed below.

Parameters present in the source region which are absent for plane wave coupling are:

- (a) Direct drive from gamma radiation
- (b) Compton current
- (c) Air conductivity
- (d)  $E_{\theta} \neq cB_{d}$
- (e) Radial electric field, Er
- (†) High field amplitudes for late times (t >  $10^{-6}$  seconds)

The impact of these additional coupling mechanisms is most severe for ranges close to the burst (say < 1000 meters). For greater ranges the impact is not so great and, in fact, at sufficiently large distances (> 2000 meters) these parameters become insignificant and the EMP

disturbance is approximately a plane wave as described in Reference 1. In this report we have selected an intermediate range. Therefore each of the phenomena listed above must be examined separately to determine its relative importance.

At ranges close to the burst, objects will be subjected to large doses of gamma radiation and Compton current. These objects may then collect charge directly from the environment. The amount of charge collected depends on the relative transparency of the various objects to gamma rays, i.e., how well gamma radiation penetrates without attenuation. An object which stops all incident radiation (opaque or  $\gamma$ -thick) will simply collect all the Compton current which strikes it, acquiring a net electric charge.

Objects which are transparent ( $\gamma$ -thin) to incident gamma radiation accumulate little net charge. Gamma rays traverse the object with essentially no attenuation causing a Compton current to flow off the back of the object. This current draws away the charge resulting from the collection of Compton current on the front. Another way to view this is to regard the gamma-thin object invisible, i.e., the Compton current passes directly through the object, or is continuous through the object interfaces. In fact, gamma thin objects may still collect about 10 percent of the incident Compton current, due to material differences.

The criterion for gamma-thick or gamma-thin is given by Longmire  $^{11}$  in terms of the projected mass of the object in the direction of the incident gamma radiation. This value of projected mass per unit area is  $m_{\gamma}=30~{\rm gm/cm^2}.$  Thus an object will be gamma-thin or gamma-thick if its projected mass per unit area is small or large compared to  $30~{\rm gm/cm^2}.$  It is expected that essentially all components of a tactical system will be relatively gamma-thin. Possible exceptions are vehicle engine blocks, armored portions of vehicles or solid propellant sections of larger missiles.

To obtain some estimate of direct effects for gamma-thick objects consider the charge accumulated on a solid sphere of 1 meter radius R. The charge is obtained approximately as the peak Compton current times the half width of the pulse. This gives

$$Q = J_c \Delta t \text{ Area} = (19 \frac{\text{amp}}{\text{m}^2})(2.5 \times 10^{-8} \text{sec}) (\pi R^2 \text{m}^2)$$

$$= 1.5 \times 10^{-6} R^2 Coul$$
.

For a radius R the capacitance of a sphere to infinity is

$$C = 1.11 \times 10^{-10} R$$
 farad

so that the potential to infinity is

$$V = \frac{Q}{C} = \frac{1.5 \times 10^{-6}}{1.11 \times 10^{-10}} R$$

becoming  $\approx$  13,500 volts for a sphere with a 1 meter radius. This might be representative of an ungrounded vehicle with an armored cab about six feet across. The electric field which corresponds to this potential (several thousand volts/meter) will affect any antennas mounted on the vehicle. In addition, a sizeable current will flow on the surface of the vehicle as surface charges form to accommodate the electric field, and these surface currents will couple into exposed cabling. Note however that direct interaction is only significant for dense objects with an appreciable cross sectional area.

For gamma-thin objects similar calculations 11 indicate small fields, and currents which are fractions of an amp. Virtually all elements of a tactical system, vans, antennas, cables, electronic equipment, etc., are expected to be gamma-thin. In situations where a question exists a separate calculation should be conducted to define the expected effects. Possible methods for simulating direct interaction in ARES will be discussed in Section 4.

The range selected in this report represents the inner edge of the tactical source region. Peak air conductivity is moderately low (~  $10^{-4}$  mho/meter) but not negligible and local gamma radiation levels are too low to cause significant charging of typical antennas through direct interaction.

### 3.2 COUPLING TO CANONICAL OBJECTS

### 3.2.1 Short Monopole Antenna

We first consider source region coupling to the short monopole. Outside of the source region the environment can frequently be characterized by a current and/or voltage source which drives an equivalent circuit representation of the antenna. The parameters in this circuit are fixed in time. In the source regin the effects of air conductivity, Compton current, volume space charge, X-rays and neutrons complicate the situation considerably.

Air conductivity is often treated in an equivalent circuit approach by introducing a time-varying resistance into the antenna equivalent circuit. The collection of Compton current is modeled by introducing an additional current source into the circuit. Results of experiments in  ${\sf AURORA}^{12}$  confirm the validity of these changes to the

normal equivalent circuit treatments of the coupling problem, at least for simple geometries.

A Thevenin equivalent circuit for the short monopole antenna is presented in Figure 3-2. Note that for the antenna to be electrically small it must not only be small compared to the wavelength of the exciting electric field, it must also be small compared to the skin depth in the ionized air produced by the burst. The skin depth in the air is given approximately by

$$\delta = \sqrt{\tau/\mu_0 \sigma} \tag{3-1}$$

where  $\tau$  is a characteristic time, and  $\mu_0$  is the permeability of free space (=  $4\pi\times10^{-7}$  henrys/meter). For the 1.2 km environment  $\tau$  =  $10^{-8}$  sec and  $\sigma$  =  $3.5\times10^{-4}$  mho/meter so that  $\delta$  ≈ 5 meters. The EMP frequency content is essentially all below 100 MHz so the wavelength is greater than 3 meters. Therefore, we may select a monopole height h = 1 meter as truly short in terms of both criteria. For antenna lengths which exceed 5 meters the ARES would excite an incorrect response. For intermediate antenna lengths, i.e., where the antenna length exceeds the exciting wavelength but not the skin depth, or vice versa, it would be necessary to determine the response on a case by case basis.

In Figure 3-2 the current source is proportional to the dose rate,  $\hat{D}$ , and represents the collection of Compton current for gamma-thick objects. The short monopole, however, is gamma-thin and the current source is absent from the circuit.

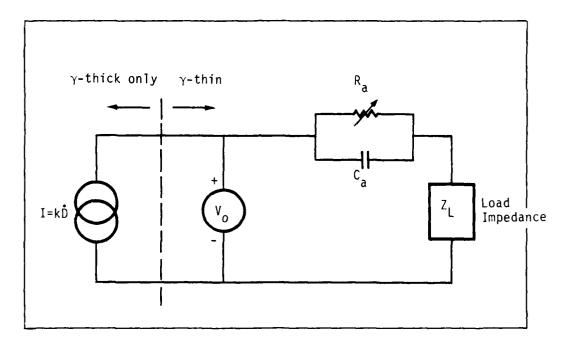


Figure 3-2. Equivalent circuit for short monopole in the source region.

The antenna capacitance is that appropriate for a thin cylinder resting on a conducting ground plane,  $^{13}$ 

$$C_{a} = 2h/cZ_{o} , \qquad (3-2)$$

where

$$Z_0 = 60 (\Omega - 3.39)$$
 (3-3)

and

$$\Omega = 2\ln(2h/a) . \tag{3-4}$$

In these expressions "h" is the height of the monopole and "a" is the radius. The shunt resistance is related to the time varying air conductivity and the antenna capacitance by

$$R_{a} = \epsilon_{o}/C_{a}\sigma , \qquad (3-5)$$

where  $\varepsilon_0(=8.854\times10^{-12}~\text{farads/m})$  is the permittivity of free space. The voltage across the antenna and its load is

$$V_0 = -\frac{1}{2} hE_x$$
 (3-6)

For h = 1 m, a = 0.01 meter we obtain  $\alpha$  = 10.59, giving C<sub>a</sub> = 15.4 pf for this monopole. The short circuit current is related to V<sub>0</sub> by

$$I_{sc} = \frac{hC_a}{2} \left[ \dot{E}_x + \frac{\sigma}{E_a} \dot{E}_x \right]$$
 (3-7)

which expresses the current in terms of environment parameters. Note that  $\epsilon_0^{\circ} \hat{\epsilon}_{\rm X}$  represents the displacement current and  $\delta \hat{\epsilon}_{\rm X}$  the conduction current. In the source region environment the two terms dominate at different times. The displacement current is dominant during the rise of the pulse with the conduction current dominant at the peak. At late times the two current densities are comparable.

The monopole response has been computed for the "typical" environment defined earlier using the MODELC computer code. A 50, load impedance was assumed as representative of a typical load. The output

from the code is the voltage across and the current through the load impedance.

The computed monopole response is presented in Figure 3-3 for a range of 1.2 km, with and without air conductivity. This comparison shows the effect of air conductivity on the coupling. The short irrelit current is presented and it is seen that air conductivity increases the peak current about 20 percent. Without air conductivity there is a large negative overshoot after the peak which is greatly reduced when conductivity is present. This behavior is easily understandable in terms of short antenna theory. Near the negative peak the displacement current and conduction current add together so conductivity enhances the response. After the peak the conduction current opposes the displacement current so the negative overshoot is reduced. In neither case is the difference very large.

For the case of no air conductivity, as in the ARES, a bipolar pulse is seen to result. For an antenna terminated in a capacitor this pulse will initially charge and then discharge the capacitor, leaving little net charge at the end of the pulse. With air conductivity present discharging does not occur and a net charge remains on the capacitor at the end of the pulse. It should be noted however that the major EMP stress on the capacitor occurs during the initial charging, and this stress is essentially the same for both cases. Thus the ARES can provide the appropriate peak stress on a major component, though this stress will not be maintained. Any system hardened to one of the waveforms would probably provide sufficient safety margin to be hard to the other. The voltage waveform shows a very similar response with the peak value increased about 15 percent with conductivity present.

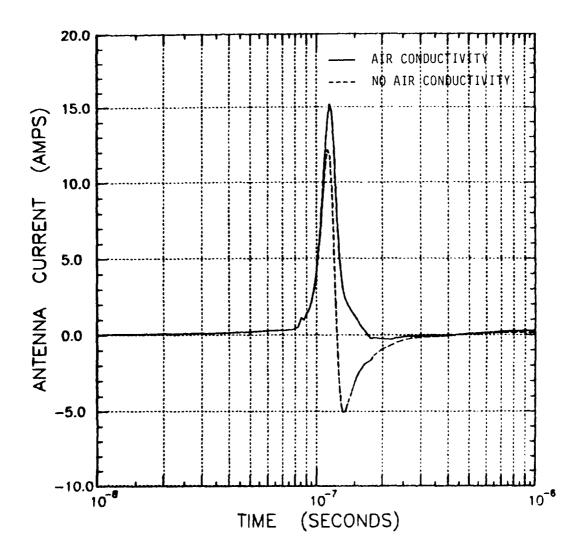


Figure 3-3. Response of a short antenna 1.2 km from a 50 kt surface burst.

For purposes of comparison similar calculations are shown in Figure 3-4 at a range of 2.0 km. Results with and without conductivity are shown. The peak current has dropped to about 2 amps. The response with and without conductivity are essentially the same so the effect of conductivity has become insignificant.

the ARES and in the source region with air conductivity. For a high impedance monopole, responding to  $\dot{E}$ , the ARES is seen to overdrive the short antenna by a factor of 6. A peak antenna current of 87 amps is obtained compared to 15 amps for the source region response. This provides ample safety margin for the test of all components with comparable areas exhibiting a dipole behavior. A negative oversnoot is also present in the ARES response since the source region dampening provided by air conductivity is not present. A negative overshoot is also present in the ARES response since the source region dampening provided by air conductivity is not present. In the same environment, a low impedance monopole would act to integrate  $\dot{E}$  so that it would respond primarily to  $\dot{E}_{\rm X}$ , the vertical electric field. For this case the ARES would overdrive the antenna about 70 percent in the region of the peak, but would substantially underdrive the monopole at late times.

### 3.2.2 Small Loop Antenna

In a deployed tactical system the myriad cables, phone lines, etc., will inevitably, together with vans, vehicles, and the conducting ground, form conducting loops. These various loops will frequently be oriented with the plane of the loop perpendicular to the ground. This is an ideal arrangement for coupling to the By component of the source region fields. The loops will in general be small for the frequencies in

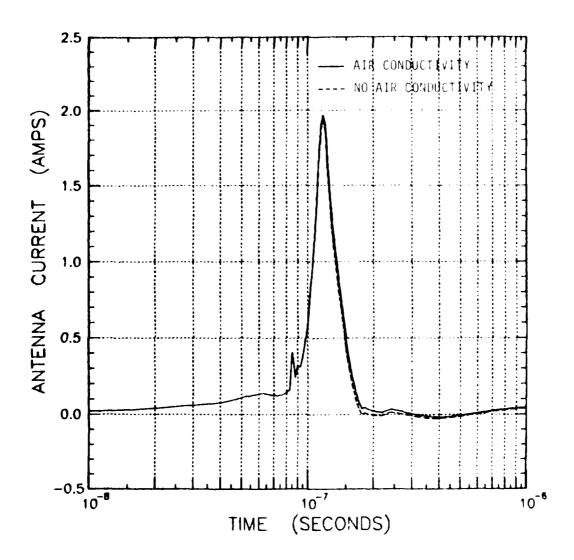


Figure 3-4. Response of a short antenna 2.0 km from a 50 kt surface burst.

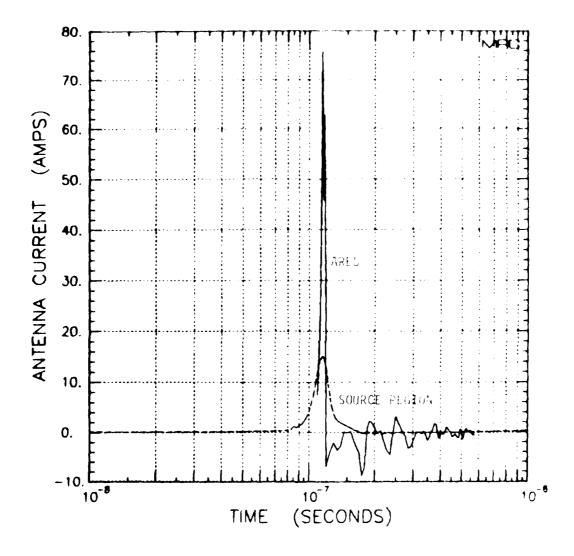


Figure 3-5. Comparison of short antenna response a Riand source region environments.

the EMP pulse. The loop must also be small compared to the plasma skin depth, z=5 meters, so that the magnetic field does not change significantly over the area of the loop.

A Inevenin equivalent circuit for a small loop is presented in Figure 3-b. Again, the current source represents the collection of Compton current for gamma-thick objects, and is absent for the case being considered. Note that the resulting equivalent circuit is the same as that obtained for plane wave coupling. The air conductivity has no impact on a loop since a magnetic dipole does not respond to changes in the dielectric constant. The corresponding parameter for a loop is the permeability  $\mu$  which in general is constant, and is usually equal to  $\mu_0$ , the free space value.

For a loop of radius R with wire radius a the inductance is (for  $\kappa \gg a)^{\frac{1}{2}}$ 

$$L_{a} = \int_{O} R_{-\kappa} n(8R/a) - 2 \text{ henrys}$$
 (3-8)

The open circuit voltage is given by

$$V_{\rm rec} \sim \beta_{\rm ph}$$

for r more all to the plane of the loop. For example, consider a loop with  $\kappa = 0.564$  s. we obtain a loop area of  $\lambda \approx 1$  ms. with a = .01 the loop indictance is  $t_{\rm d} = s \cdot 1$  shorty. The open circuit voltage for the loop is given by readtion set which expresses the antenna response in terms of the relevant the field parameter, by, the time changing magnetic field.

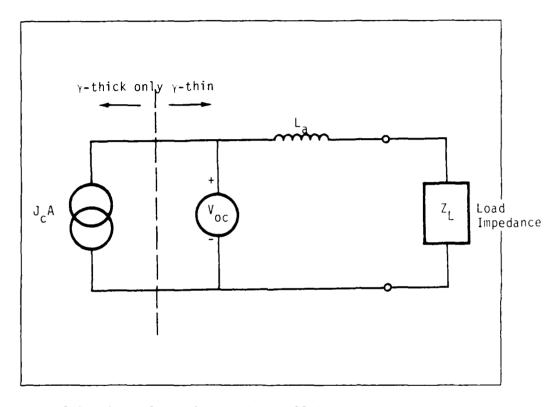


Figure 3-6. Equivalent circuit for small loop in the source region.

Figure 3-7 shows the response of a small loop in the ARES and in the source region. Since the loop area was conveniently selected as 1  $\rm m^2$  this figure appears identical to the  $\rm B_y$  time derivative shown in

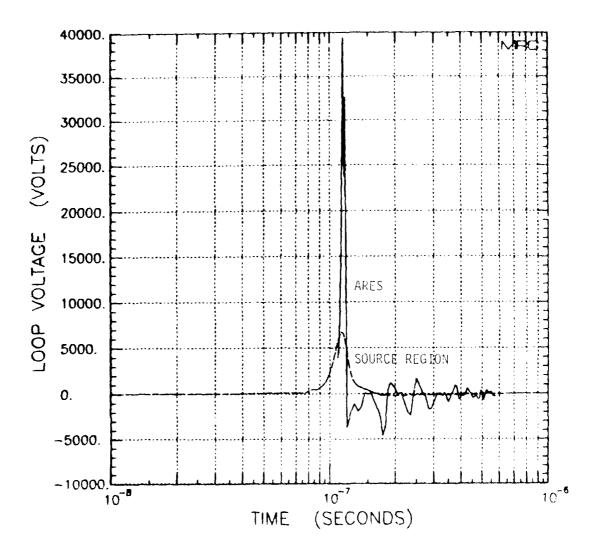


Figure 3-7. Comparison of small loop response in ARES and source region environments.

Figure 2-12. For a high impedance loop the response is to  $\dot{b}_y$  and the peak amplitude in the ARES is 39,000 volts compared to 7000 volts in the source region, so the loop is overdriven by about 6 times. However, a low impedance loop integrates  $\dot{b}_y$  so that it responds essentially to B. In this case the ARES would provide a valid test at the peak of the pulse but would underdrive the loop substantially at late times.

# SECTION 4 ARES MODIFICATIONS FOR SOURCE REGION TESTS

The purpose of this effort is not to redesign the ARES, but rather to estimate the quality of a source region simulation in ARES. It appears, however, that some improvement in this simulation might be achieved by installing some simple, and reversible, modifications in the ARES facility. It is important that any changes remain reversible since the ARES shall be required as a plane wave simulator for the MX program in 1983 and 1986. Also, some modifications may be accomplished on the equipment under test rather than on the test facility. The following paragraphs describe the important concept of simulation fidelity and provide some specific suggestions for improving the fidelity of a source region simulatin in the ARES.

Simulation fidelity is a measure of the degree to which the simulator produces a response in a test object that is relatable to the threat response. From this viewpoint there is a complementary relation between simulator realism on the one hand and theoretical understanding with calculational ability on the other. In one limit a totally realistic eproduction of the threat environment would produce a representative response, and no understanding is required (except to relate the results to another threat). At the other limit, if it is possible to calculate the response with adequate confidence, no test is required. The key to evaluating simulation fidelity is an appreciation of what is being

simulated, and what assumptions are required to relate the results to a threat exposure.

 $\,$  As discussed in previous sections the phenomena that distinguish SREMP from free-field EMP are:  $^{14}$ 

- (1) Air conductivity (time-varying).
- (2) Compton currents, with material interface effects.
- (3) Ionization effects in electronics, whereby the impedance presented to input/output wires is changed.

The most realistic simulation would produce the correct time-dependent gamma dose rate, with a spectrum that produces the correct absorption coefficient and Compton-current densities. If only the air conductivity is important, the requirement on the spectrum can be relaxed to producing the correct dose-rate history in air only. If the time dependence of the air conductivity is not critical the rise time of the ionization source can be increased, so that the dose rate at the time of interest (when the EM field is applied) is correct. If the relation between dose rate and air conductivity is completely unimportant, the air conductivity could be simulated by immersing the test object in a volume conducting medium (e.g., conducting foam). At this stage, one can also conceive of simulating the net Compton currents by adding appropriate high-impedance pulsed current sources connected directly to the test object.

This discussion illustrates the basic concept of simulation fidelity: a tradeoff between simulation realism and understanding. Each element of understanding simplifies the simulation problem, either by making a parameter irrelevant or producible by simpler simulation.

The definition of the source region adopted for this report permits a significant relaxation of simulator requirements for testing. It has been seen that pulse risetime and peak field values in the ARES already provide a quite good representation of the source region. Late time histories are not duplicated but neither are they as important in many cases. Source region simulation in the ARES may be further improved by adopting one or more of the following techniques:

- (a) Soil on the ARES floor
- (b) Conducting foam
- (c) Auxiliary pulsers
- (d) Current injection
- (e) Direct cable drive

None of these methods involve any direct modification of the ARES facility and only (a) requires an extended time frame (at least several months) to justify the effort involved.

A simple technique for obtaining a more realistic simulation of ground effects involves covering the ARES floor space with soil to a depth of about 4 meters. Elements of the tactical system are then deployed on the soil and tested with an actual air-ground interface. Presence of the soil would also serve to enhance the horizontal electric field  $(E_Z)$  giving a higher peak value more in line with the computed radial field. Also, if the soil surface is inclined in the z-direction from front to rear, cable drives will be enhanced since they will also be driven by a portion of the  $E_X$  field component. Remapping of the fields would be required in the ARES working space to validate values which are predicted beforehand. The movement of this large quantity of soil involves a substantial effort but is not terribly significant with a time frame of

years between installation and removal. The floor ground plane may be protected by plastic sheets.

Systems with electronic components will be affected by the photons and neutrons present in the source region. These effects need to be included in an overall coupling analysis since TREE effects can affect the coupling of the system to the environment by, for example, modifying the input impedance. Possible synergisms between EMP and TREE effects are another area where the source region is different from the free-field environment. If pre-test predictions indicate that certain system elements are particularly sensitive to radiation, a local ionization source may be introduced.

Other circumstances which call for local radiation involve instances where air conductivity adjacent to an object or inside a cavity produce a significant modification of the object response. It would then be desirable to reproduce the local air conductivity in the larger system simulation.

If the time dependence of the air conductivity is unimportant it may be simulated by immersing the test object in a volume conducting medium such as a conducting foam. Again, this technique is best suited for simulations on local objects such as an antenna. Immersing an entire vehicle in foam may be possible but is probably not practical. There will be many instances where air conductivity will be unimportant and its simulation may be dispensed with altogether.

Supplementary field excitation of particular objects might be required for some simulations. Auxiliary pulsars might be employed to enhance or modify the field environment provided by the ARES. Such units

represent common technology but would require a special installation for each test. It has been suggested that an auxiliary pulser may be employed to simulate the late time fields present in the source region. This pulser would possess a slow rise time and would tend to "fill in" at the end of the ARES pulse.

As previously discussed some components of the deployed system could be gamma-thick, e.g., an armored vehicle. These components will incur a net electric charge in the source region environment resulting from direct interaction. This direct interaction may be simulated by current injection directly onto the object of interest. Sources and feed lines could be buried in the soil with minimum affect on the simulation. Pulse risetimes and peak amplitudes are readily within the state-of-the-art.

Another instance where direct drive might be useful is the accumulation of current for long cable runs. Simulation of currents induced on long runs associated with particular equipments may be accomplished by allowing the cable to exit the ARES working volume. Direct drive might then be accomplished with equipment exterior to the ARES facility.

lesting of internal interaction efects on system components such as vans cannot be accomplished in the ARES and must be handled in other simulators, such as AURORA, which are designed to produce an X-radiation environment. Some related work has already been performed. Testing in other simulators will serve to compliment the ARES to provide a more thorough evaluation of the entire system response.

## SECTION 5 SUMMARY AND CONCLUSIONS

The previous sections have defined a tactical source region envionment and made comparisons with the environment provided by the ARES. Coupling of both EMP environments into specified canonical objects have been examined. In the short time span available for the current study this effort could not be exhaustive but some interesting results have been obtained. The effort presented above has emphasized EMP coupling into deployed tactical systems since the ARES appears particularly suited to the evaluation of large system components. As such, this effort has only addressed the inner edge of the tactical source region, but this represents the important region where crew survival is possible (and therefore where equipment survival is necessary).

A summary of pros and cons for the ARES as a source region simulator is presented in Table 5-1. In general it is concluded that those aspects of the source region environment which are absent in the ARES are either unimportant or may be provided on an individual basis for specific system elements. It has been shown that many important coupling elements (dipoles, loops and transmission lines) respond primarily to time derivatives of the impinging EMP,  $\tilde{E}_z$ ,  $\tilde{E}_x$  and  $\tilde{B}_y$ . The ARES provides a significant overtest capability (about 6 times) for these time derivatives.

Table 5-1. Source region simulation using the ARES.

COMMENTS	Large size is unique Restoration of facility to original configuration is straightforward	Various supplementary sources would be required to provide absent parameters for particular objects and tests.	for dipole and loop responses
\$\$\tag{00}	l. Outdoor facility - weather dependent	2. Absence of some significant aspects of source region environment  a) radiation & neutrons b) Compton current c) air conductivity d) late time high field amplitudes	
17 A. V.	Separional  Little Little Labable of test- replatuable Ruise Little Foretter turn-around, ruiting anoth periday  Little Control Long Can be as the firm a una region	Fig. 1940 Fig. Therefore of source (F. a. Tree edge of source F. 3137. F. peak field arbitudes Correct field components present with correct sense (Frequency spectrum	e) i and o exceed source millor values

Table 5-1 (continued).

COMMENTS

CONS

PROS

er Fields also not uniform in source region; not a problem if fields are known	S	Environment a strong function of crew survival (rather than of device characteristics per se)	Requires placement of earth on working volume floor
3. Fields not uniform over entire working volume	4. Frequencies not tested = 17 MHz plus multiples		
		6. Can test wide range of devices	7. Air ground interface can be reasonably represented

6. Cannot test effects on unmanned system components located closer to burst

5. Power consumed for lingering source region pulse not represented The important air-ground interface may be included in ARES tests by the simple expedient of covering the ARES working floor space with earth. Where air conductivity is deemed important it may be provided by using conducting foams around particular system elements. Direct current injection might also be employed on some objects to supplement the environment provided by the ARES. It appears that any modifications necessary to supplement the ARES environment may be accomplished directly on the test objects or within an earth covered floor space without impact on the ARES facility proper.

Recycling time for the ARES pulser is only three minutes while the entire facility may be recycled in about five minutes. This provides a significant advantage when compared with such source region facilities as AURORA which require much longer recycle times (2 hours). Costs for both facilities appear comparable, but the ARES could provide considerably more data for a given test period.

It is suggested that further study be performed to provide firm recommendations for an ARES source region test facility. A specific system should be selected and examined in detail. This approach would result in firm suggestions for modifications to supplement the environment provided by the ARES. In particular, it is necessary to examine the effect of soil on the ARES floor space.

In closing, we firmly agree with the following quote from the ARES facility User's Manual  $^{1/2}$ :

"The data obtained through experimentation at ARES are necessarily only a portion of the information eventually required for an assessment of the EMP vulnerability and survivability on any system. The ARES Facility is not can stand alone, unsupported by analysis and other test efforts. As a general rule, a comprehensive assessment program should be well under way before an attempt is made to design the experiment to be conducted at ARES.

Such an assessment program will include identification and ranking by relative importance of all paths of entry of electromagnetic energy into the weapon system, identification of the critical circuits within the system, determination of the environment parameters requiring close reproduction during test and analysis of the coupling paths within the system. Clearly, some sest effort will have to be completed prior to designing a test program to be executed at the ARES. The preliminary testing, such as experimental verification of circuit thresholds, should be completed and understood before the ARES testing is begun.

Regardless of the type of system being tested, the yield of the experimental effort will depend on the care with which the experiment is designed and executed, especially with regard to the characteristics of the facility and the ways in which these characteristics satisfy the user's EMP requirements."

It appears that the ARES will prove to be a useful source region simulator for tactical Army systems when proper attention is focussed on the design of the test. There is merit to testing systems with small antennas in the ARES. These test results are test relatable. It is possible to test both low and high altitude susceptability of tactical systems simultaneously, with the ARES capable of providing an overtest of many important system components.

#### REFERENCES

- 1. Sponn, D.J., private communication.
- 2. Berberet, J.A. and A. Lowrey, The Approach of an EMP Hardened Tactical Army, System EMP Hardening Seminar, NOSC, San Diego, California, 7-9 August 1979.
- 3. Glasstone, S. and P.J. Dolan, eds., <u>The Effects of Nuclear Weapons</u>, 3rd Edition, U.S. Government Printing Office, 1977.
- 4. Longmire, C.L., On the Electromagnetic Pulse Produced by Nuclear Explosions, IEEE Trans. on Antennas and Propagation, Vol. AP-26, No. 1, January 1978.
- 5. Crevier, W.F., and E.D. Kalasky, Evaluation of AURURA as a Tactical Source Region Simulator, AURURA EMP Memo 21, Mission Research Corporation, Santa Barbara, California.
- 6. Wirth, J.C., ARES Facility Description, Facilities, Equipment and Operation, Volume I, Final Report, DNA 3265F-1, Braddock, Dunn and McDonald, Inc., Albuquerque, New Mexico, 12 March 1974.
- 7. Hutchins, R.L., ARES Facility Description, Electromagnetic Environment, Volume II, Final Report, DNA 3265F-2, Braddock, Dunn and McDdonald, Inc., Albuquerque, New Mexico, 12 March 1974.
- 8. Dyche, J.W. and R.A. Kitter, ARES Facility Description, Data Processing System, Volume III, Final Report, DNA 3265F-3, Braddock, Dunn and McDonald, Inc., Albuquerque, New Mexico, 12 March 1974.
- 9. Preliminary, Tall Pines Field Mapping Report, BUM/IAC-79-063-IR, The BUM Corporation, Albuquerque, New Mexico, March 1979.
- 10. Evans, R.D., <u>The Atomic Nucleus</u>, McGraw-Hill Book Co., New York, 1955.

- 11. Longmire, C.L., Direct Interaction Effects in EMP, Lake-R-12, Los Alamos Anclear Comporation, January 1971; also Interaction note na, Canuary 1971.
- 12. Merkle, G., D.J. Sponn, C.L. Longnire and L.F. Crevier, An Equivalent Circuit Analysis of the HDL Concentric Cylinder Experiments in AURORA, IEEE Trans. Nucl. Sci., Vol. NS-24, No. 6, December 1977.
- 13. Messiar, M.A., Coupling to Antennas, Chapter 10, Revised, DAA Handbook, Mission Research Corporation, Santa Barbara, CA.
- 14. Baum, C.E., EMP Simulators for Various Types of Muclear EMP Environments: An Interio Categorization, IEEE Trans. Ant. and Prop. Vol. AP-26, No. 1, January 1978.
- 15. V.A.J. van Lint, private communication.
- 10. Meyer, U.L., W.L. Chadsey and V.J. Pine, Internal Salvin Response of a Communications Van to AURORA Irradiation, Intel Trans. Nucl. Sci., Vol. NS-24, No. 6, December 1977.
- 17. ARES User's Manual, BDM/FAC-78-103-FR, The BDM Comporation, Albuquerque, New Mexico, May 1978.

### DISTRIBUTION LIST

transca legh widevs en las Affer Beasse M-Ms. Schools Affer Beasse M-Ms. Affer Beasse Mark Affer Beasse M-Ms
ATTM: DELINER, A. John B. Gradensea entagent of the Army ATTM: December 19 of the Army ATTM: DECember 19 of the ATTM: DECember 11 ATTM: DELMB-N-H ATTM: DELMB-
ry Diamond Laterate re- arithmet of the Arry Allin Diamonth TI Allin Dilmonth F Allin Dilmonth Ti Allin Diamonth Ti Allin Diamonth Mas
ATM: BLECHAM ATM: BLECHAM ATM: BLECHAM ATM: BLECHAM ATM: BLECHAM ATM: BLECHAM ATM: BLECHAM
ATMS SEARCHAN ATMS SEARCHANN CHRONIC CONTROL ATMS SEARCHANN
Johnson (1997) in the service of the second of the service of the second
CHERT CARREST CONTROL OF CONTROL OF CONTROL OF CARREST CONTROL OF
j Admilija, sa de duatri kilosisisisisisisisisi ATTND SA MAA ATTND SA MARILT
. Any place on the common of a second of the common of the
<ul> <li>Array to present this to be the control of the contro</li></ul>
Amus Teta (Trade) a visa - Amus Teta (Trade) a visa - ATMs (Teta Net) a visa a visa
Charles Marker (et al., 1988) (1988) (1988) Charles Marker (et al., 1988) (1988) (1988) Charles Marker (et al., 1988) (1988) (1988)

 $(\sqrt{1}\log(29)\sqrt{1}) = 1 + \log(2\sqrt{1}\log(2\sqrt{1}))$ SEPARAMENT OF THE ARMS THE CONTROL OF Charming and patchare and La la Maria de Caracteria de la Caracteria de Caracteria d white is to Mossias wants grown to the Artivity of the Artivit Maskilo edul pranti Chi califolita et Maskilo en Chi califolita et Maskilo en Chi califolita Chi califolita Chi califolita Chi califolita CHARTMAN BUT E NOVE New Control Steps Connected States with South News of that the partial tension to be the Districted by H Notice of the first parties are made of a management of the first parties of the first partie Nave Charles Christopher Chyria (1944) - Bartoner Stiffe (1951) - Love Collection generalise de la companya de la comp La companya de No. Communication (Communication Communication) The second secon Navigation of the state of the Security of th Mass. The control of The second secon Mariel Dittor Weimer Getter 1996 - Graffel Distribute 1998 - German German 1998 - German Distribute Oracle Degrada (Editor)
 Oracle Degrada (Editor)
 Oracle Degrada (Editor)
 Oracle Degrada (Editor) Tales of the control of the field meeriquentains Strain (1999) had sever fairly (1996) had (1997) 677 Mg (1907) No. 1 Leading to the Advantage State (1982) Sendinguar for the party of the first of the American Control of the first of the ACM and the first of the fi Some Adelia and Demand selectors And Connect of the Connected Addition to the where the computation was a constitution of the  $\mathcal{D}_{\rm s}$ fraters Archeman coate of the object ATML MID INFO Service ATML MID INFO SERVICE ATML MACHINES

DEPARTMENT OF LETERAL FOR 1867 TO SERVICE GOLD MAY OF IMPROV Charles Stars Origin Costs D. Albert J. North, Albert M. W. 94 persections of Energy All June no operation Office All Mines and Child All Colleges American file to may be a comdepartment of therak Teller Affice of Stillit, System, 1. Starill Consider the section of  $\mathcal{M}_{\mathrm{sec}}$  , where  $\mathcal{M}_{\mathrm{sec}}$  , where THE R. GRADIENMAN, ALLEGED S. Sayater 1 for eller of Starten et al. Contral Intol. Lamber Adems v ATTAL OSAK MED Communities to definish ordation ACTAN Case Discover one gehanisang relati Afthronication professional control for the control of the control second frem republic Carrier, ATTAL (STEM, Martha  $\underset{i\in\mathcal{I}}{\text{Theorem }} \mathcal{A}_{i} = \{ (i,j) \in \mathcal{I}_{i}, \dots, (i,j) \in \mathcal{I}_{i} \}$ La Calotter No. 10 of the scale of ONT Rose TORS. Estation Tennes 1933 le 2015 de la companya della companya de la companya della c Consider Designational Laboration of the Consideration of the Considerat Edwin wash Abalistical as a second of the con-First Arresse ( A Contest to Martin Contest Armin [] Parametrical Armin () extreme Los Arakos Millional Editentifo Lab Millo We esta I. Costani Altin Me essa, J. Matti Altin L. Feston Ford Appropriate Schemidt, Latitle Const. Charling to testate with the F. Thompson credit Nathonal Rab office R. Parker Alln: I. Martin Alln: J. Cours perental constants of a Attach Research ( Denata LEGARIMENT OF DEFENDE COUTRACTORS General Great School of Attraction Research Little for ikeM Corp. Afthr I Abrary Approximation to the Control of Approximation (Control of Control factorial Clestric Co. All Mr. The Milyanson Specific ter. All'Si M. Frank General Hestern to CATTALL TOO MERCAL COMMEN Computed Co.

AT By D. Yemle

ATTR: W. domes

PYM: next be beload fibrary
ATTR: B. damrahan Cursas — H.MPo Allho A. McCarlotta Allho A. Rathertona Allho (A. Rathertona Social withits to weller Harman S. M. C. Addition of the Co. to, Allow of carolton, loc Arther A. during willow located at a toraty accounts to the first terminate of a second care Alloper Comp. Alloper Compression

#### DEPARTMENT OF DEFENSE CONTRACTORS (Continued)

Georgia Institute of Technology ATIN: Res & Sec Coord for H. Denny

Grumman Aerospace Corp ATIN: L-01 35

Harris Corporation

ATTN: A. Strain ATTN: V. Pres & Min Proms Div

mazeltime Corp ATTN: J. Ukrent

noneywell, Inc ATIN: R. Johnson ATIN: SWRC Library

noneywell, inc AITN: W. stewart AITN: S. Graff

magnes Airuraft Co ATTN: E. walker ATTN: J. Singletary ATTN: CTOC 6/E110

pagne, Almonaft Co AlTN: A. Namevsky

III Research Institute

ATTN: J. Bridges ATTN: I. Mindel

Institute for Defense Analyses

ATTA: Tech Info Services

International Tel & Telegraph Corp ATTN: Technical Library

13% Physics corp ATTM: H. Milde ATTM: R. Evans

191 Corp ATTN: N. Rudie

ATTY: b. williams

Alin: w. Radusky

JARCON

" ATTN: H. Stabl ATTN: E. Wessas

TAYOUR

AlTh: Library

Raman Sciences Corp ATIN: W. ware ATIN: E. shelton ATIN: A. shidges ATIN: W. Rich ATIN: J. Lubell ATIV: N. Beauchamp

.:tton Systems, Inc. ATTY: MO4 -61 ATTY: (MC-GP

DEPARTMENT OF DEFENSE CONTRACTORS (Continued)

Litton Systems, Inc. ATIN: J. Skaggs

Lockheed Missiles & Space Co., Inc.

ATIN: Technical Information Center

Lockneed Missiles & Space Co, Inc

ATTN: E. Smith ATTN: L. Rossi ATTN: G. Heath

ATIN: n. Theyn 2 cy ATIN: S. Talmuty

M.I.T. Lincoln Lab ATTN: L. Loughlin

Martin Marietta Corp. ATTN: M. Griftin

Martin Marietta Corp ATTN: G. Freyer

McDonnell Douglas Corp

ATTN: I. Enger

McDonnell Douglas Corp.

ATTN: Technical Library Services ATTN: S. Schneider

Mission Research Corp

ATTN: EMP Group ATTN: A. Hart ATTN: D. Holst 5 cy ATTN: Document Costrol

Mission Research Corp ATTN: A. Chodorow ATTN: L. McCormick

Mission Research Corp. San Diego.

ATIN: V. Van Lint

Mission Research Corporation

ATIN: Sec Off for w. Stark

MITRL Corp.

ATTN: M. Fitzgerald

Norden Systems, Inc

ATTN: Technical Library

Northrop Corp ATIN: V. Demartino ATIN: B. Ahlport

ATIN: Lew Smith

Pacific-Sierra Research Corp ATTN: H. Brode

Palisades Inst for Roch Service, Inc.

ATTN: Records Supervisor

Physics International Co.

AllN: Document Control

#### DEPARTMENT OF DEFENSE CONTRACTORS (Continued)

R & D Associates

ATTN: Document Control ATTN: R. Schafer

ATTN: B. Gage

ATTN: P. Haas

R & D Associates ATTN: J. Bombardt

Rane Corp

ATTN: LIE-D

Raytheon Co

ATTN: G. Joshi

Raytheon Co

ATTN: H. Flescher

RCA Corp ATTN: G. Brucker

RCA Corp ATTN: L. Minich ATTN: D. O'Connor

Rockwell International Corp ATTN: V. Michel ATTN: 0/243-060, 031-CA31

ATTN: J. Monroe

Rockwell International Corp

ATTN: B. White

Rockwell International Corp

ATTN: B-1 DIV TIC, BAOB

Rockwell International Corp ATTN: F. Shaw

Science Applications, Inc.

ATTN: R. Parkinson

Science Applications, Inc ATTN: W. Chadsey

Sidney Frankel & Associates

ATTN: S. Frankel

Singer Co

ATTN: Technical Information Center

perry Rand Corp ATTN: Technical Library

AVCU Research & systems Group ATTN: Library As30

Battelle Memorial Institute ATTN: E. Leach

BOM Corp ATTN: Corporate Library

DEPARTMENT OF DEFENSE CONTRACTORS (Continued.

Spenny Rand Comp ATTN: D. Schow

Spire Corp

ATT's: R. Little

SRI International

ATTN: A. Whitson ATTN: B. Gaster ATTN: E. Vance

Sylvania tystems Group ATTN: L. Elaisdell ATTN: C. Thornmill

.vivaria Systems Group AITN: J. waidron AITN: A. Novenski AITN: E. Motthok AITN: J. Flood AITN: M. Nuretons AITN: (. Ramshotton

Systems, kienke & Software, Inc. ATTN: A. wilson

Teledyne Brown Engineering ALTN: F. Leonard

Tokas Instruments, Inc. AT(N: Technica) (Holary ATTN: D. Manu.

Texas Tech Criver its ATTM: T. Simpson

TRW Defense & Space Sys Group ATTY: e. Holloway ATTY: J. Adams ATTY: L. Madnolia

JRw Detense & space Sys Group ATTN: P. Montenser

Varian Associates, Inc

ATIN: H. Jory

Westinghouse Electric Corp 67.3ct Technical Library

Acrospace Corp.
ATTW: E. Feerleter
ATTW: E. Feerleter
ATTW: P. Colta.
ATTW: P. Colta.
ATTW: J. Amediate.
ATTW: J. Feerla Co.

Asharias Albertates ATTN: Charact

